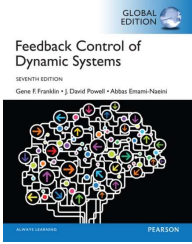

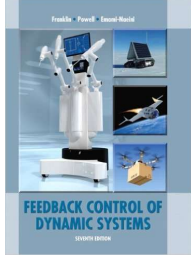


Feedback Control of System *Prof. Cheng-Hsien Liu*






國立清華大學
National Tsing Hua University



PME320702

Feedback Control of Systems



Chapter 3

Instructor : Cheng-Hsien Liu 劉承賢

liuch@pme.nthu.edu.tw

<http://mx.nthu.edu.tw/~chhslu>

Lecture 3 1

1

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Chapter 1: An Overview and Brief History of Feedback Control

Chapter 2: Dynamic Models

→ **Chapter 3: Dynamic Response**

Chapter 4: A First Analysis of Feedback

Chapter 5: The Root-locus Design Method

Chapter 6: The Frequency-response Design Method

Chapter 7: State-space Design

Chapter 8: Digital Control

Chapter 9: Nonlinear Systems

Chapter 10: Control Systems Design: Principles and Case Studies

Appendix A: Laplace Transforms

Appendix B: Solutions to the Review Questions

Appendix C: MATLAB Commands

Appendix WA: A Review of Complex Variables

Appendix WB: Summary of Matrix Theory

Appendix WC: Controllability and Observability

Appendix WD: Ackermann's Formula for Pole Placement

Appendix W2.1.4: Complex Mechanical Systems

Lecture 3 2

2

Chapter 3

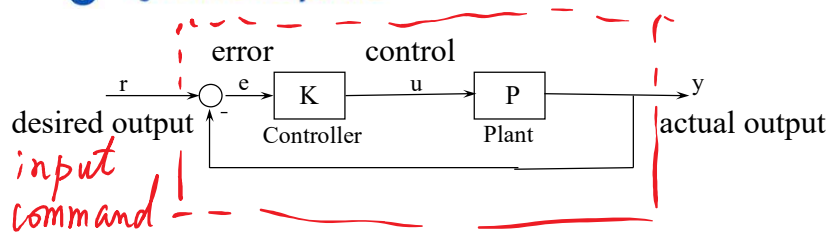
Dynamic Response

Lecture 3

3

3

3 Dynamic Response



3.1 Review of Laplace Transforms ←

3.2.1 The Block Diagram

3.3 Effect of Pole Locations

3.4 Time-Domain Specifications

3.5 Effects of Zeros and Additional Poles

3.6.3 Routh's Stability Criterion

Lecture 3

4

4

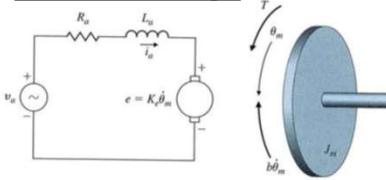
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3.1 Review of Laplace Transforms

Linear Time Invariant Dynamic System

Mechatronic Example



↗ **2nd order linear dynamic system**

$$J_m \ddot{\theta}_m + b \dot{\theta}_m = K_t i_a \quad (2.52)$$

$$L_a \frac{di_a}{dt} + R_a i_a = v_a - K_e \dot{\theta}_m \quad (2.53)$$

↘ **1st order linear dynamic system**

n 階 系統

For a general nth order linear dynamic system, represented by parameter variable

$$\begin{cases} \frac{d^n y(t)}{dt^n} + a_{n-1} \frac{d^{n-1} y(t)}{dt^{n-1}} + a_{n-2} \frac{d^{n-2} y(t)}{dt^{n-2}} + \dots + a_1 \frac{dy(t)}{dt} + a_0 y(t) \\ = b_m \frac{d^m f(t)}{dt^m} + b_{m-1} \frac{d^{m-1} f(t)}{dt^{m-1}} + \dots + b_1 \frac{df(t)}{dt} + b_0 f(t) \end{cases}$$

the coefficients $a_i, i = 0, 1, 2, \dots, n - 1$, and $b_i, i = 0, 1, 2, \dots, m$, are assumed to be constant, which indicates that the given system is time invariant.

Lecture 3 5

5

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Linear Dynamic System

For a general nth order linear dynamic system, represented by parameter variable

$$\begin{cases} \frac{d^n y(t)}{dt^n} + a_{n-1} \frac{d^{n-1} y(t)}{dt^{n-1}} + a_{n-2} \frac{d^{n-2} y(t)}{dt^{n-2}} + \dots + a_1 \frac{dy(t)}{dt} + a_0 y(t) \\ = b_m \frac{d^m f(t)}{dt^m} + b_{m-1} \frac{d^{m-1} f(t)}{dt^{m-1}} + \dots + b_1 \frac{df(t)}{dt} + b_0 f(t) \end{cases} \Rightarrow \text{線性方程式}$$

y: 因變數
t: 自變數

Linear Time variant Dynamic System

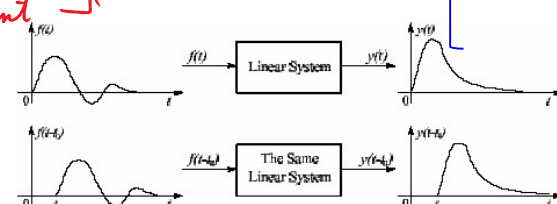
$a_i(t) \rightarrow$

Linear Time Invariant Dynamic System

$a_i = \text{constant} \rightarrow$

\Rightarrow 因變數不自乘 (1)
不自乘 (2)
不在 sin, cos, log (內) (3)

↗ **LTI Dynamic System**



Graphical representation of system time invariance

Lecture 3 6

6

Transfer Functions

轉移函數

LTI Dynamic System

$$\begin{aligned} \Rightarrow T.F. \quad G(s) &= \frac{Y(s)}{F(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \cdots + b_1 s + b_0}{s^n + a_{n-1} s^{n-1} + \cdots + a_1 s + a_0} \\ &= K \frac{(s + z_1)(s + z_2) \cdots (s + z_m)}{(s + p_1)(s + p_2) \cdots (s + p_n)} \end{aligned}$$

Def.:

1. $G(s)$ is said to be **proper** if $G(\infty)$ is a finite constant $\rightarrow n \geq m$
2. $G(s)$ is said to be **strictly proper** if $G(\infty) = 0 \rightarrow n > m$
3. **Relative order** : $n-m$
4. The **order of system**: n
5. **Poles**: $-p_1, -p_2, \dots, -p_n$ 極點
6. **Zeros**: $-z_1, -z_2, \dots, -z_n$ 零點

Note: $\left\{ \begin{array}{l} \text{T.F. is to describe the relationship between input and output} \\ \text{T.F. is independent to the input and initial condition} \end{array} \right.$

The \mathcal{L}_- Laplace Transform

Inverse Laplace Transform

Laplace transform of $f(t)$, denoted by $\mathcal{L}_-\{f(t)\} = F(s) \quad s = \sigma + j\omega,$

$$\Rightarrow \left\{ \begin{array}{l} F(s) \triangleq \int_{0^-}^{\infty} f(t) e^{-st} dt, \end{array} \right. \quad (3.24) \quad \text{Laplace}$$

$$\left. \begin{array}{l} f(t) = \frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F(s) e^{st} ds, \end{array} \right\} \quad (3.25) \quad \text{inverse Laplace}$$

The \mathcal{L} - Laplace Transform

Inverse Laplace Transform

TABLE A.1

Properties of Laplace Transforms

Number	Laplace Transform	Time Function	Comment
—	$F(s)$	$f(t)$	Transform pair
1	$\alpha F_1(s) + \beta F_2(s)$	$\alpha f_1(t) + \beta f_2(t)$	Superposition
2	$F(s)e^{-s\lambda}$	$f(t - \lambda)$	Time delay ($\lambda \geq 0$)
3	$\frac{1}{ a } F\left(\frac{s}{a}\right)$	$f(at)$	Time scaling
4	$F(s + a)$	$e^{-at}f(t)$	Shift in frequency
5	$s^m F(s) - s^{m-1}f(0) - s^{m-2}\dot{f}(0) - \dots - f^{(m-1)}(0)$	$f^{(m)}(t)$	Differentiation
6	$\frac{1}{s} F(s)$	$\int_0^t f(\zeta) d\zeta$	Integration
7	$F_1(s)F_2(s)$	$f_1(t) * f_2(t)$	Convolution
8	$\lim_{s \rightarrow \infty} sF(s)$	$f(0^+)$	Initial Value Theorem
9	$\lim_{s \rightarrow 0} sF(s)$	$\lim_{t \rightarrow \infty} f(t)$	Final Value Theorem
10	$\frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F_1(\zeta)F_2(s - \zeta) d\zeta$	$f_1(t)f_2(t)$	Time product
11	$\frac{1}{2\pi} \int_{-j\infty}^{+j\infty} Y(-j\omega)U(j\omega) d\omega$	$\int_0^\infty y(t)u(t) dt$	Parseval's Theorem
12	$-\frac{d}{ds} F(s)$	$tf(t)$	Multiplication by time

TABLE A.2

Table of Laplace Transforms

Number	$F(s)$	$f(t), t \geq 0$
1	1	$\delta(t)$
2	$1/s$	$1(t)$
3	$1/s^2$	t
4	$2!/s^3$	t^2
5	$3!/s^4$	t^3
6	$m!/s^{m+1}$	t^m
7	$\frac{1}{s+a}$	e^{-at}
8	$\frac{1}{(s+a)^2}$	te^{-at}
9	$\frac{1}{(s+a)^3}$	$\frac{1}{2!}t^2e^{-at}$
10	$\frac{1}{(s+a)^m}$	$\frac{1}{(m-1)!}t^{m-1}e^{-at}$
11	$\frac{a}{s(s+a)}$	$1 - e^{-at}$
12	$\frac{a}{s^2(s+a)}$	$\frac{1}{a}(at - 1 + e^{-at})$
13	$\frac{b-a}{(s+a)(s+b)}$	$e^{-at} - e^{-bt}$
14	$\frac{s}{(s+a)^2}$	$(1-at)e^{-at}$
15	$\frac{a}{s(s+a)^2}$	$1 - e^{-at}(1+at)$
16	$\frac{(b-a)s}{(s+a)(s+b)}$	$be^{-bt} - ae^{-at}$
17	$\frac{s}{s^2+a^2}$	$\sin at$
18	$\frac{s}{s^2+a^2}$	$\cos at$
19	$\frac{s+a}{(s+a)^2+b^2}$	$e^{-at} \cos bt$
20	$\frac{b}{(s+a)^2+b^2}$	$e^{-at} \sin bt$
21	$\frac{a^2+b^2}{s[(s+a)^2+b^2]}$	$1 - e^{-at} \left(\cos bt + \frac{a}{b} \sin bt \right)$

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The \mathcal{L}_- Laplace Transform

Laplace transform of $f(t)$, denoted by $\mathcal{L}_-\{f(t)\} = F(s)$ $s = \sigma + j\omega_s$

$$F(s) \triangleq \int_{0^-}^{\infty} f(t)e^{-st} dt. \quad (3.24) \quad \text{Laplace}$$

Inverse Laplace Transform

$$f(t) = \frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F(s)e^{st} ds, \quad (3.25) \quad \text{inverse Laplace}$$

Lecture 3- Review Laplace 11

11

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Joseph-Louis Lagrange
Italian 1736-1813

Pierre-Simon Laplace
French 1749-1827

Any periodic function can be rewritten as a weighted sum of Sines and Cosines of different frequencies 任何函數都可以展開為三角級數

Fourier Transform - Review

Time Domain \longleftrightarrow Frequency Domain

$$\hat{f}(\xi) = \int_{-\infty}^{\infty} f(x) e^{-2\pi i \xi x} dx$$

Fourier transform for functions f

$$f(x) = \sum_{n=-\infty}^{\infty} c_n e^{2\pi i(n/T)x} = \sum_{n=-\infty}^{\infty} \hat{f}(\xi_n) e^{2\pi i \xi_n x} \Delta \xi,$$

$$\begin{cases} c_n = \frac{1}{T} \int_{-T/2}^{T/2} f(x) e^{-2\pi i(n/T)x} dx. \\ c_n = (1/T) \hat{f}(n/T) \end{cases}$$

Lecture 3- Review Laplace 12

12

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$F(s) \triangleq \int_{0^-}^{\infty} f(t)e^{-st} dt$ *Laplace*

$f(t) = \frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F(s)e^{st} ds$ *inverse Laplace*

$s = j2\pi\xi$

Fourier Transform

$F(\xi) = \int_{-\infty}^{\infty} f(x)e^{-j2\pi\xi x} dx$

$f(x) = \int_{-\infty}^{\infty} F(\xi)e^{j2\pi\xi x} d\xi$

Fourier Transform

$$\hat{f}(w) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} f(x)e^{-iwx} dx$$

$$f(x) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \hat{f}(w)e^{iwx} dw$$

→ Laplace transform of $f(t)$, denoted by $\mathcal{L}\{f(t)\} = F(s)$ $s = \sigma + j\omega$

Lecture 3- Review Laplace 13

13

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The \mathcal{L} - Laplace Transform

Laplace transform of $f(t)$, denoted by $\mathcal{L}\{f(t)\} = F(s)$ $s = \sigma + j\omega$

Inverse Laplace Transform

→ $F(s) \triangleq \int_{0^-}^{\infty} f(t)e^{-st} dt$ (3.24) *Laplace*

$f(t) = \frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F(s)e^{st} ds$ (3.25) *inverse Laplace*

Lecture 3- Review Laplace 14

14

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View I

$$\begin{cases} M\ddot{x} + b\dot{x} = 0.63i, \\ L\frac{di}{dt} + Ri = v_a - 0.63\dot{x} \end{cases} \Rightarrow \begin{cases} 2\ddot{x} + 3\dot{x} = 0.63i \\ 4\dot{i} + 5i = 6\sin(2t) - 0.63\dot{x} \end{cases}$$

$$\Rightarrow \begin{cases} 2s^2X + 3sX = 0.63I \\ 4sI + 5I = 6\frac{2}{s^2 + 4} - 0.63X \end{cases} \Rightarrow$$

View II

$F_c d + M_D = I\ddot{\theta}$
 $U = F_c d + M_D$ $\Rightarrow I\ddot{\theta} = U \Rightarrow \frac{\Theta(s)}{U(s)} = \frac{1}{I s^2}$

Lecture 3- Review Laplace 15

15

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Def: 1

EXAMPLE 3.7 Impulse Function Transform

$$p_{\Delta}(t) = \begin{cases} \frac{1}{\Delta}, & 0 \leq t \leq \Delta \\ 0, & \text{elsewhere} \end{cases} \quad (3.3)$$

$$\lim_{\Delta \rightarrow 0} p_{\Delta}(t) = \delta(t), \quad \text{impulse.} \quad (3.5) \quad \delta(t) = 0 \quad t \neq 0, \quad (3.8)$$

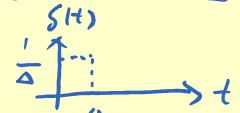
$\delta(t-5)$

$$\Rightarrow \int_{-\infty}^{\infty} \delta(t) dt = 1. \quad (3.9)$$

$$\Rightarrow \int_{-\infty}^{\infty} f(\tau) \delta(t - \tau) d\tau = f(t). \quad (3.10)$$

Lecture 3- Review Laplace 16

16

$$\odot \int_{-\infty}^{\infty} \delta(t) dt = \lim_{\Delta \rightarrow 0} \int_0^{\Delta} \frac{1}{\Delta} dt = \lim_{\Delta \rightarrow 0} 1 = 1$$


A graph showing a rectangular pulse function $\delta(t)$ on a coordinate system. The horizontal axis is labeled t and the vertical axis is labeled $\delta(t)$. The pulse has a constant height of $\frac{1}{\Delta}$ and a width of Δ , extending from $t=0$ to $t=\Delta$. Dashed lines indicate the height and width of the pulse.

$$\odot \int_{-\infty}^{\infty} f(\tau) \delta(t-\tau) d\tau = f(t)$$

$$\delta(t) = \begin{cases} \frac{1}{\Delta} & 0 < t < \Delta \\ 0 & \text{elsewhere} \end{cases}$$

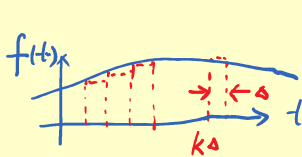
$$\delta(t-k\Delta) = \begin{cases} \frac{1}{\Delta} & k\Delta < t < (k+1)\Delta \\ 0 & \text{elsewhere} \end{cases}$$

Lecture 3- Review Laplace

17

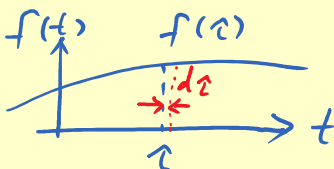
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Proof:



$$\hat{f}(t) = \sum_{k=-\infty}^{\infty} f(k\Delta) \delta(t-k\Delta) \cdot \Delta$$

($\delta(t-k\Delta) \cdot \Delta = 1$)



$$f(t) = \int_{-\infty}^{\infty} f(\tau) \delta(t-\tau) d\tau$$

Lecture 3- Review Laplace

18

18

$$\delta(t) \Rightarrow \textcircled{1} \int_{-\infty}^{\infty} \delta(t-a) dt = 1$$

$$\Rightarrow \textcircled{2} \int_b^c \delta(t-a) dt = \begin{cases} 1 & a \in [b, c] \\ 0 & a \notin [b, c] \end{cases}$$

$$\Rightarrow \textcircled{3} \int_{-\infty}^{\infty} \delta(t-a) f(t) dt = f(a)$$

$$\Rightarrow \textcircled{4} \int_b^c \delta(t-a) f(t) dt = \begin{cases} f(a) & a \in [b, c] \\ 0 & a \notin [b, c] \end{cases}$$

$$\mathcal{L}[\delta(t)] = \int_0^{\infty} \delta(t) e^{-st} dt$$

$$\Rightarrow \int_{-\infty}^{\infty} f(\tau) \delta(t-\tau) d\tau = f(t)$$

$$\int_{-\infty}^{\infty} f(\tau) \delta(\tau-t) d\tau = f(t)$$

$$= \int_0^{\infty} e^{-s\tau} \delta(\tau-0) d\tau$$

$$= e^{-s \cdot 0} = 1$$

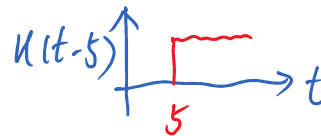
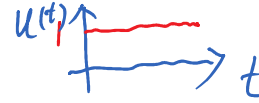
$$\Rightarrow \mathcal{L}[\delta(t)] = 1$$

Def 2 Unit step function $u(t) \leftrightarrow 1(t)$

$$u_s(t) = \begin{cases} 1 & t \geq 0 \\ 0 & t < 0 \end{cases}$$

$$\Rightarrow u_s(t-a) = \begin{cases} 1 & t \geq a \\ 0 & t < a \end{cases}$$

$$\Rightarrow \delta(t) = \frac{d}{dt} u_s(t)$$



$$\begin{aligned} \textcircled{1} f(t) = u(t) &= \begin{cases} 0 & t < 0 \\ 1 & t \geq 0 \end{cases} & \mathcal{L}[u(t)] &= \mathcal{L}[1] \\ & & &= \int_0^{\infty} e^{-st} dt \\ & & &= -\frac{e^{-st}}{s} \Big|_0^{\infty} = \frac{1}{s} \leftarrow \end{aligned}$$

② Unit function 奧運較功能

$$f(t) \uparrow \begin{array}{c} \text{Graph of } f(t) = \sin(t-a) \times u(t-a) \end{array} \rightarrow t \Rightarrow f(t) = \sin(t-a) \times u(t-a)$$

$$f(t) \uparrow \begin{array}{c} \text{Graph of } f(t) = u(t-a) - u(t-b) \end{array} \rightarrow t \Rightarrow f(t) = u(t-a) - u(t-b)$$

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$$u'(t) = \delta(t)$$

$$\Rightarrow \begin{cases} \mathcal{L}[u(t)] = \frac{1}{s} \\ \mathcal{L}[\delta(t)] = \int_0^{\infty} e^{-st} \delta(t) dt = e^{-st} \Big|_0 = 1 \end{cases}$$

$$\Rightarrow \begin{cases} \mathcal{L}^{-1}[1] = \delta(t) \\ \mathcal{L}^{-1}\left[\frac{1}{s}\right] = u(t) \end{cases} \quad \begin{matrix} \swarrow \text{diff} \\ \searrow \text{diff} \end{matrix} \quad \left\{ \begin{array}{l} \mathcal{L}^{-1}[s] = \delta'(t) \\ \mathcal{L}^{-1}[s^n] = \delta^n(t) \end{array} \right.$$

$$\Rightarrow \mathcal{L}[\delta(t-a)] = e^{-sa}$$

Lecture 3- Review Laplace 23

23

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$$\int_{-\infty}^{\infty} \delta'(t) f(t) dt = \delta(t) \checkmark f(t) \Big|_{-\infty}^{\infty} - \int_{-\infty}^{\infty} \delta(t) f'(t) dt = (-1) f'(t) \Big|_{t=0}$$

$$\Rightarrow \int_{-\infty}^{\infty} \delta^{(n)}(t-a) f(t) dt = (-1)^n f^{(n)}(t) \Big|_{t=a}$$

Lecture 3- Review Laplace 24

24

$$\begin{aligned}
 \mathcal{L}[t^a] &= \int_0^{\infty} e^{-st} t^a dt \\
 &\begin{cases} \text{Let } st = y \Rightarrow dt = dy/s \\ \downarrow \end{cases} \\
 &= \int_0^{\infty} \frac{y^a}{s^a} e^{-y} \frac{dy}{s} = \frac{1}{s^{a+1}} \int_0^{\infty} y^a e^{-y} dy \\
 &= \frac{1}{s^{a+1}} \left[-y^a e^{-y} \Big|_0^{\infty} + (a) \int_0^{\infty} y^{a-1} e^{-y} dy \right] \\
 &= \frac{a}{s^{a+1}} \int_0^{\infty} y^{a-1} e^{-y} dy \\
 &= \dots = \frac{a!}{s^{a+1}} \int_0^{\infty} e^{-y} dy
 \end{aligned}$$

$$\begin{aligned}
 &= \frac{a!}{s^{a+1}} \left[-e^{-y} \Big|_0^{\infty} \right] \\
 &= \frac{a!}{s^{a+1}} [-0 + 1] = \frac{a!}{s^{a+1}} \quad \text{16}
 \end{aligned}$$

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$$\begin{cases} \mathcal{L}[\sin \omega t] \\ \mathcal{L}[\cos \omega t] \end{cases} \begin{matrix} \searrow \\ \rightarrow \end{matrix} e^{i\omega t} = \cos \omega t + i \sin \omega t$$

$$\Rightarrow \mathcal{L}[e^{i\omega t}] = \int_0^{\infty} [e^{i\omega t} e^{-st}] dt$$

$$= \int_0^{\infty} e^{(i\omega - s)t} dt = \frac{1}{i\omega - s} e^{(i\omega - s)t} \Big|_0^{\infty}$$

$$= \dots = \frac{s + i\omega}{s^2 + \omega^2}$$

$$\Rightarrow \begin{cases} \mathcal{L}[\sin \omega t] = \frac{\omega}{s^2 + \omega^2} \quad \text{背} \\ \mathcal{L}[\cos \omega t] = \frac{s}{s^2 + \omega^2} \quad \text{背} \end{cases}$$

Lecture 3- Review Laplace 27

27

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$$\mathcal{L}[e^{at}] = \int_0^{\infty} e^{at} e^{-st} dt = \frac{-e^{-(s-a)t}}{s-a} \Big|_0^{\infty}$$

$$= \frac{1}{s-a} \quad \text{背}$$

Lecture 3- Review Laplace 28

28

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TABLE A.2 **Table of Laplace Transforms** **Mentioned So far!**

Number	$F(s)$	$f(t), t \geq 0$	
1	1	$\delta(t)$	← p. 20
2	$1/s$	$1(t)$	← p. 22/ p. 28
3	$1/s^2$	t	← p. 25-26
4	$2!/s^3$	t^2	← p. 25-26
5	$3!/s^4$	t^3	← p. 25-26
6	$m!/s^{m+1}$	t^m	← p. 25-26
7	$\frac{1}{s+a}$	e^{-at}	← p. 28
8	$\frac{1}{(s+a)^2}$	te^{-at}	
9	$\frac{1}{(s+a)^3}$	$\frac{1}{2!}t^2e^{-at}$	
10	$\frac{1}{(s+a)^m}$	$\frac{1}{(m-1)!}t^{m-1}e^{-at}$	
11	$\frac{a}{s(s+a)}$	$1 - e^{-at}$	
12	$\frac{a}{s^2(s+a)}$	$\frac{1}{a}(at - 1 + e^{-at})$	
13	$\frac{b-a}{(s+a)(s+b)}$	$e^{-at} - e^{-bt}$	
14	$\frac{a^2}{(s+a)^2}$	$(1-at)e^{-at}$	
15	$\frac{a^2}{s(s+a)^2}$	$1 - e^{-at}(1+at)$	
16	$\frac{(b-a)s}{(s+a)(s+b)}$	$be^{-bt} - ae^{-at}$	
17	$\frac{a}{s^2+a^2}$	$\sin at$	← p. 27
18	$\frac{s}{s^2+a^2}$	$\cos at$	← p. 27
19	$\frac{s+a}{(s+a)^2+b^2}$	$e^{-at} \cos bt$	
20	$\frac{b}{(s+a)^2+b^2}$	$e^{-at} \sin bt$	
21	$\frac{a^2+b^2}{s[(s+a)^2+b^2]}$	$1 - e^{-at} \left(\cos bt + \frac{a}{b} \sin bt \right)$	

Lecture 3- Review Laplace 29

29

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Appendix A Laplace Transforms

4. Shift in Frequency

Multiplication (modulation) of $f(t)$ by an exponential expression in the time domain corresponds to a shift in frequency:

$$F_1(s) = \int_0^\infty e^{-at} f(t) e^{-st} dt = \int_0^\infty f(t) e^{-(s+a)t} dt = F(s+a). \quad (A.9)$$

Lecture 3- Review Laplace 30

30

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TABLE A.2 **Table of Laplace Transforms** **Mentioned So far!**

Number	F(s)	f(t), t ≥ 0	
1	1	δ(t)	← p. 20
2	1/s	1(t)	← p. 22/ p. 28
3	1/s ²	t	← p. 25-26
4	2!/s ³	t ²	← p. 25-26
5	3!/s ⁴	t ³	← p. 25-26
6	m!/s ^{m+1}	t ^m	← p. 25-26
7	$\frac{1}{s+a}$	e ^{-at}	← p. 28
8	$\frac{1}{(s+a)^2}$	te ^{-at}	← p. 28 + p. 25-26
9	$\frac{1}{(s+a)^3}$	$\frac{1}{2!}t^2e^{-at}$	← p. 28 + p. 25-26
10	$\frac{1}{(s+a)^m}$	$\frac{1}{(m-1)!}t^{m-1}e^{-at}$	← p. 28 + p. 25-26
11	$\frac{a}{s(s+a)}$	1 - e ^{-at}	
12	$\frac{a}{s^2(s+a)}$	$\frac{1}{a}(at - 1 + e^{-at})$	
13	$\frac{b-a}{(s+a)(s+b)}$	e ^{-at} - e ^{-bt}	
14	$\frac{a^2}{(s+a)^2}$	(1-at)e ^{-at}	
15	$\frac{s}{s(s+a)^2}$	1 - e ^{-at} (1+at)	
16	$\frac{(b-a)s}{(s+a)(s+b)}$	be ^{-bt} - ae ^{-at}	
17	$\frac{a}{s^2+a^2}$	sin at	← p. 27
18	$\frac{s}{s^2+a^2}$	cos at	← p. 27
19	$\frac{s+a}{(s+a)^2+b^2}$	e ^{-at} cos bt	← p. 28 + p. 27
20	$\frac{b}{(s+a)^2+b^2}$	e ^{-at} sin bt	← p. 28 + p. 27
21	$\frac{a^2+b^2}{s[(s+a)^2+b^2]}$	1 - e ^{-at} (cos bt + $\frac{a}{b}$ sin bt)	

Lecture 3- Review Laplace **31**

31

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The \mathcal{L}_- Laplace Transform Laplace transform of $f(t)$, denoted by $\mathcal{L}_-\{f(t)\} = F(s)$

$$F(s) \triangleq \int_{0^-}^{\infty} f(t)e^{-st} dt$$

Inverse Laplace Transform

$$f(t) = \frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F(s)e^{st} ds$$

⇒ $\mathcal{L}\{y'(t)\} = \int_0^{\infty} y'(t)e^{-st} dt = [y(t)e^{-st}]_0^{\infty} - \int_0^{\infty} y(t)e^{-st}(-s) dt$

$$= 0 - y(0) + s \int_0^{\infty} y(t)e^{-st} dt$$

$$= sY(s) - y(0)$$

⇒ $\mathcal{L}\left\{\frac{df}{dt}\right\} = \int_{0^-}^{\infty} \left(\frac{df}{dt}\right)e^{-st} dt = -f(0^-) + sF(s)$

⇒ $\mathcal{L}\{\ddot{f}\} = s^2F(s) - sf(0^-) - \dot{f}(0^-)$

⇒ $\mathcal{L}\{f^{(m)}(t)\} = s^mF(s) - s^{m-1}f(0^-) - s^{m-2}\dot{f}(0^-) - \dots - f^{(m-1)}(0^-)$

Lecture 3- Review Laplace **32**

32

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Convolution

$\mathcal{L}\{f_1(t)\} = F_1(s)$ and $\mathcal{L}\{f_2(t)\} = F_2(s)$

$\Rightarrow \mathcal{L}\{f_1(t) * f_2(t)\} = \int_0^\infty f_1(t) * f_2(t) e^{-st} dt = \int_0^\infty \left[\int_0^t f_1(\tau) f_2(t - \tau) d\tau \right] e^{-st} dt$

$= \int_0^\infty \int_\tau^\infty f_1(\tau) f_2(t - \tau) e^{-st} dt d\tau$

Multiplying by $e^{-s\tau} e^{s\tau}$

$= \int_0^\infty f_1(\tau) e^{-s\tau} \left[\int_\tau^\infty f_2(t - \tau) e^{-s(t-\tau)} dt \right] d\tau$

change variables $t' \triangleq t - \tau$

$= \int_0^\infty f_1(\tau) e^{-s\tau} d\tau \int_0^\infty f_2(t') e^{-st'} dt' = F_1(s) F_2(s)$

$\Rightarrow \mathcal{L}^{-1}\{F_1(s) F_2(s)\} = f_1(t) * f_2(t)$

Lecture 3- Review Laplace 33

33

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Convolution (My viewpoint)

$S^2 Y + Y = F$

物理系統

方程式 $\begin{cases} \ddot{y} + y = f(t) \\ F \cdot \frac{1}{s^2+1} = Y \end{cases}$

Initial Condition $\begin{cases} y(t=0) = 0 \\ y'(t=0) = 0 \end{cases}$

$\Rightarrow f(t) * g(t) = y(t)$

convolution

Lecture 3- Review Laplace 34

34

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$$\ddot{y} + y = \delta(t)$$

↓ Laplace

$$s^2 Y + Y = 1$$

$$Y = \frac{1}{s^2 + 1}$$

↓ Laplace⁻¹

$$y = \sin t$$

$$\mathcal{L}[\delta(t)] = \int_0^\infty \delta(t) e^{-st} dt$$

$$\Rightarrow \int_{-\infty}^\infty f(\tau) \delta(t - \tau) d\tau = f(t)$$

$$\int_{-\infty}^\infty f(\tau) \delta(\tau - t) d\tau = f(t)$$

$$= \int_0^\infty e^{-s\tau} \delta(\tau - 0) d\tau$$

$$= e^{-s \cdot 0} = 1$$

$$\Rightarrow \mathcal{L}[\delta(t)] = 1$$

Lecture 3- Review Laplace 35

35

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$$\ddot{y} + y = f(t)$$

⇒ Assume $f(t) = \delta(t) \iff y(t) = g(t) = \sin t$

$f_1(t) = \delta(t)$

\Rightarrow
 $y_1(t) = \sin t$

$f_2(t) = \delta(t - \Delta t)$
 $= 2\delta(t - \Delta t)$

\Rightarrow
 $y_2(t) = \sin(t - \Delta t)$

$f_3(t) = \delta(t - 2\Delta t)$

\Rightarrow
 $y_3(t) = \sin(t - 2\Delta t)$

物理系统

$\ddot{y} + y = \delta$
 ↓ Laplace
 $s^2 Y + Y = 1$
 $Y = \frac{1}{s^2 + 1}$
 ↓ Laplace⁻¹
 $y = \sin t$

Lecture 3- Review Laplace 36

36

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Assume $\Delta t \rightarrow 0$
 $\Rightarrow f(t) = \sum_{n=0}^{(t/\Delta t)-1} f(n\Delta t)$

$\Rightarrow y(t) = f(0) \sin t + f(\Delta t) \sin(t-\Delta t) + f(2\Delta t) \sin(t-2\Delta t) + \dots$

$$= \sum_{n=0}^{N-1} f(n\Delta t) \sin(t-n\Delta t)$$

$$= \sum_{n=0}^{N-1} f(n\Delta t) \cdot g(t-n\Delta t)$$

as $\Delta t \rightarrow 0$

$$\Rightarrow y(t) = \int_0^t f(\tau) g(t-\tau) d\tau$$

$$= f(t) * g(t) \quad \text{convolution}$$

物理系统

$$\sum_{k=1}^m \dots$$

Laplace

$$\sum Y + Y = 1$$

$$Y = \frac{1}{s+1}$$

Laplace

$$y = \sin t$$

Lecture 3- Review Laplace 37

37

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Convolution

$$f_1(t) * f_2(t) = \int_0^t f_1(\tau) f_2(t-\tau) d\tau$$

Lecture 3- Review Laplace 38

38

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Convolution

$$\Rightarrow f_1(t) * f_2(t) = \int_0^t f_1(\tau) f_2(t - \tau) d\tau$$

Lecture 3- Review Laplace 39

39

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Convolution

$$\mathcal{L}\{f_1(t)\} = F_1(s) \text{ and } \mathcal{L}\{f_2(t)\} = F_2(s)$$

$$\Rightarrow \mathcal{L}\{f_1(t) * f_2(t)\} = \int_0^\infty f_1(t) * f_2(t) e^{-st} dt = \int_0^\infty \left[\int_0^t f_1(\tau) f_2(t - \tau) d\tau \right] e^{-st} dt$$

$$= \int_0^\infty \int_\tau^\infty f_1(\tau) f_2(t - \tau) e^{-st} dt d\tau$$

Multiplying by $e^{-s\tau} e^{s\tau}$

$$= \int_0^\infty f_1(\tau) e^{-s\tau} \left[\int_\tau^\infty f_2(t - \tau) e^{-s(t-\tau)} dt \right] d\tau$$

change variables $t' \triangleq t - \tau$

$$= \int_0^\infty f_1(\tau) e^{-s\tau} d\tau \int_0^\infty f_2(t') e^{-st'} dt' = F_1(s) F_2(s)$$

$$\Rightarrow \mathcal{L}^{-1}\{F_1(s) F_2(s)\} = f_1(t) * f_2(t)$$

Lecture 3- Review Laplace 40

40

Inverse Laplace Transform by Partial-Fraction Expansion

 $m < n$

$$F(s) = \frac{b_1 s^m + b_2 s^{m-1} + \dots + b_{m+1}}{s^n + a_1 s^{n-1} + \dots + a_n} = \frac{C_1}{s-p_1} + \frac{C_2}{s-p_2} + \dots + \frac{C_n}{s-p_n}. \quad (3.43)$$

$$(s-p_1)F(s) = C_1 + \frac{s-p_1}{s-p_2}C_2 + \dots + \frac{(s-p_1)C_n}{s-p_n}. \quad (3.44)$$

$$C_1 = (s-p_1)F(s)|_{s=p_1} \quad C_i = (s-p_i)F(s)|_{s=p_i}$$

EXAMPLE 3.9

$$Y(s) = \frac{(s+2)(s+4)}{s(s+1)(s+3)} \quad \text{Find } y(t)$$

$$Y(s) = \frac{C_1}{s} + \frac{C_2}{s+1} + \frac{C_3}{s+3}$$

$$\left[\begin{array}{l} C_1 = \frac{(s+2)(s+4)}{(s+1)(s+3)} \Big|_{s=0} = \frac{8}{3} \\ C_2 = \frac{(s+2)(s+4)}{s(s+3)} \Big|_{s=-1} = -\frac{3}{2} \\ C_3 = \frac{(s+2)(s+4)}{s(s+1)} \Big|_{s=-3} = -\frac{1}{6} \end{array} \right. \Rightarrow Y = \frac{8}{3} + \frac{(-\frac{3}{2})}{s+1} + \frac{(-\frac{1}{6})}{s+3}$$

$$\Rightarrow y(t) = \frac{8}{3}1(t) - \frac{3}{2}e^{-t}1(t) - \frac{1}{6}e^{-3t}1(t)$$

Lecture 3- Review Laplace

41

41

$$\text{Ex } F(s) = \frac{s^5 + 1}{(s+1)(s+2)(s+3)^3}$$

$$= 1 + \left(\frac{A}{s+1} + \frac{B}{s+2} + \frac{C}{(s+3)^3} + \frac{D}{(s+3)^2} + \frac{E}{s+3} \right)$$

⇒ 解 A, B, C, D, E.

$$\left\{ \begin{array}{l} A = \lim_{s \rightarrow (-1)} F \cdot (s+1) \\ B = \lim_{s \rightarrow (-2)} F \cdot (s+2) \\ C = \lim_{s \rightarrow (-3)} F \cdot (s+3)^3 = \lim_{s \rightarrow (-3)} \left[(s+3)^3 + A \frac{(s+3)^3}{s+1} + B \frac{(s+3)^3}{s+2} + C + D(s+3) + E(s+3)^2 \right] \end{array} \right.$$

Lecture 3- Review Laplace

42

42

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$$\begin{cases} D = \lim_{s \rightarrow (-3)} \frac{1}{1!} \frac{d}{ds} [F \cdot (s+3)^3] \\ E = \lim_{s \rightarrow (-3)} \frac{1}{2!} \frac{d}{ds^2} [F \cdot (s+3)^3] \end{cases}$$

for E only $\Rightarrow \lim_{s \rightarrow \infty} s \cdot [F(s) - 1]$

$$= \lim_{s \rightarrow \infty} s \cdot \frac{s^5 + 1 - (s^5 + 12s^4 + 56s^3 + 126s^2 + 135s + 54)}{(s+1)(s+2)(s+3)^3}$$

$$= -12 = A + B + \textcircled{E}$$

Lecture 3- Review Laplace 43

43

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Summary

<p>The \mathcal{L}- Laplace Transform</p> <p>Laplace transform of $f(t)$, denoted by $\mathcal{L}\{f(t)\} = F(s)$ $s = \sigma + j\omega$,</p> $F(s) \triangleq \int_{0^-}^{\infty} f(t)e^{-st} dt. \quad (3.24)$	<p>Inverse Laplace Transform</p> $f(t) = \frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F(s)e^{st} ds, \quad (3.25)$
----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------	----------------------------------------------------------------------------------------------------------------------------------------------

背 Basic properties of impulse (δ) function and step (u) function!

Lecture 3- Review Laplace 44

44

$$\begin{cases} \mathcal{L}^{-1}[1] = \delta(t) \\ \mathcal{L}^{-1}\left[\frac{1}{s}\right] = u(t) \end{cases} \quad \leftarrow \begin{cases} \mathcal{L}^{-1}[s] = \delta'(t) \\ \mathcal{L}^{-1}[s^n] = \delta^n(t) \end{cases}$$

$$\begin{aligned} \mathcal{L}[t^a] &= \int_0^{\infty} e^{-st} t^a dt \\ &\quad \left(\begin{array}{l} \text{Let } st = y \Rightarrow dt = dy/s \\ \text{Then } \int_0^{\infty} \frac{y^a}{s^{a+1}} e^{-y} \frac{dy}{s} = \frac{1}{s^{a+1}} \int_0^{\infty} y^a e^{-y} dy \end{array} \right) \\ &= \frac{1}{s^{a+1}} \left[-y^a e^{-y} \Big|_0^{\infty} + (a) \int_0^{\infty} y^{a-1} e^{-y} dy \right] \\ &= \frac{a}{s^{a+1}} \int_0^{\infty} y^{a-1} e^{-y} dy \\ &= \dots = \frac{a!}{s^{a+1}} \int_0^{\infty} e^{-y} dy \\ &= \frac{a!}{s^{a+1}} \left[-e^{-y} \Big|_0^{\infty} \right] \\ &= \frac{a!}{s^{a+1}} [-0 + 1] = \frac{a!}{s^{a+1}} \end{aligned}$$

$$\Rightarrow \begin{cases} \mathcal{L}[\sin \omega t] = \frac{\omega}{s^2 + \omega^2} \\ \mathcal{L}[\cos \omega t] = \frac{s}{s^2 + \omega^2} \end{cases}$$

$$\begin{aligned} \mathcal{L}[e^{at}] &= \int_0^{\infty} e^{at} e^{-st} dt = \frac{-e^{-(s-a)t}}{s-a} \Big|_0^{\infty} \\ &= \frac{1}{s-a} \end{aligned}$$

The Final Value Theorem 终值定理

背

If all poles of $sY(s)$ are in the left half of the s -plane, then

$$\Rightarrow \lim_{t \rightarrow \infty} y(t) = \lim_{s \rightarrow 0} sY(s). \quad (3.46)$$

EXAMPLE 3.10

$$Y(s) = \frac{3(s+2)}{s(s^2+2s+10)}$$

$$y(\infty) = sY(s)|_{s=0} = \frac{3 \cdot 2}{10} = 0.6.$$

A.1.3 The Initial Value Theorem 起始值定理

背

For any Laplace transform pair,

$$\Rightarrow \lim_{s \rightarrow \infty} sF(s) = f(0^+). \quad (A.27)$$

Proof: Textbook pp 836 (A.1.3)~838 (A.1.4)

Homework

Partial Expansion 整理

<1> 先化為真分式

<2> 分解後，一次分母項先解

$$<3> \frac{A}{(s+a)^m} + \frac{B}{(s+a)^{m-1}} + \dots + \frac{Z}{(s+a)}$$

<i> 消去 $(s+a)^m$ 後代 $(-a)$ 解 A

<ii> 代極限值 $sF(s)|_{s \rightarrow \infty}$ 解 Z

$$<iii> B = \frac{1}{1!} \frac{d}{ds} [\dots] \quad \downarrow \text{真分式}$$

$$C = \frac{1}{2!} \frac{d^2}{ds^2} [\dots]$$

$$<4> \text{分母二次式} \quad \frac{ps+q}{(s+a)^2+b^2} = \frac{p}{(s+a)^2+b^2} + \frac{m \cdot b}{(s+a)^2+b^2}$$

$$\mathcal{L}^{-1} \Rightarrow p e^{-at} \cos bt + m e^{-at} \sin bt$$

The \mathcal{L} - Laplace Transform

Inverse Laplace Transform

TABLE A.1

Properties of Laplace Transforms			Mentioned so far
Number	Laplace Transform	Time Function	Comment
—	$F(s)$	$f(t)$	Transform pair ←
1	$\alpha F_1(s) + \beta F_2(s)$	$\alpha f_1(t) + \beta f_2(t)$	Superposition ← p. 47
2	$F(s)e^{-s\lambda}$	$f(t - \lambda)$	Time delay ($\lambda \geq 0$) ← p. 47
3	$\frac{1}{ a } F\left(\frac{s}{a}\right)$	$f(at)$	Time scaling ← p. 47
4	$F(s + a)$	$e^{-at}f(t)$	Shift in frequency ← p. 47
5	$s^m F(s) - s^{m-1}f(0) - s^{m-2}\dot{f}(0) - \dots - f^{(m-1)}(0)$	$f^{(m)}(t)$	Differentiation ← p. 47
6	$\frac{1}{s} F(s)$	$\int_0^t f(\zeta) d\zeta$	Integration ← p. 47
7	$F_1(s)F_2(s)$	$f_1(t) * f_2(t)$	Convolution ← p. 33~40
8	$\lim_{s \rightarrow \infty} sF(s)$	$f(0^+)$	Initial Value Theorem ← p. 49
9	$\lim_{s \rightarrow 0} sF(s)$	$\lim_{t \rightarrow \infty} f(t)$	Final Value Theorem ← p. 49
10	$\frac{1}{2\pi j} \int_{\sigma_c - j\infty}^{\sigma_c + j\infty} F_1(\zeta)F_2(s - \zeta) d\zeta$	$f_1(t)f_2(t)$	Time product
11	$\frac{1}{2\pi} \int_{-j\infty}^{+j\infty} Y(-j\omega)U(j\omega) d\omega$	$\int_0^\infty y(t)u(t) dt$	Parseval's Theorem
12	$-\frac{d}{ds} F(s)$	$tf(t)$	Multiplication by time ← p. 47

TABLE A.2

Table of Laplace Transforms

Number	$F(s)$	$f(t), t \geq 0$	Mentioned so far
1	1	$\delta(t)$	←
2	1/s	1(t)	←
3	1/s ²	t	←
4	2!/s ³	t ²	←
5	3!/s ⁴	t ³	←
6	m!/s ^{m+1}	t ^m	←
7	$\frac{1}{s+a}$	e^{-at}	←
8	$\frac{1}{(s+a)^2}$	te^{-at}	←
9	$\frac{1}{(s+a)^3}$	$\frac{1}{2!}t^2e^{-at}$	←
10	$\frac{1}{(s+a)^m}$	$\frac{1}{(m-1)!}t^{m-1}e^{-at}$	←
11	$\frac{a}{s(s+a)}$	$1 - e^{-at}$	←
12	$\frac{a}{s^2(s+a)}$	$\frac{1}{a}(at - 1 + e^{-at})$	←
13	$\frac{b-a}{(s+a)(s+b)}$	$e^{-at} - e^{-bt}$	←
14	$\frac{s}{(s+a)^2}$	$(1-at)e^{-at}$	←
15	$\frac{s+a}{(s+a)^2}$	$1 - e^{-at}(1+at)$	←
16	$\frac{(s+a)s}{(s+a)(s+b)}$	$be^{-bt} - ae^{-at}$	←
17	$\frac{a}{s^2+a^2}$	$\sin at$	←
18	$\frac{s}{s^2+a^2}$	$\cos at$	←
19	$\frac{b}{(s+a)^2+b^2}$	$e^{-at} \cos bt$	←
20	$\frac{a}{(s+a)^2+b^2}$	$e^{-at} \sin bt$	←
21	$\frac{a^2+b^2}{s[(s+a)^2+b^2]}$	$1 - e^{-at} \left(\cos bt + \frac{a}{b} \sin bt \right)$	←

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EXAMPLE 3.14 Forced Differential Equation Solution

$$\ddot{y}(t) + 5\dot{y}(t) + 4y(t) = 3, \text{ where } y(0) = \alpha, \quad \dot{y}(0) = \beta$$

$$s^2 Y(s) - s\alpha - \beta + 5[sY(s) - \alpha] + 4Y(s) = \frac{3}{s} \quad Y(s) = \frac{s(s\alpha + \beta + 5\alpha) + 3}{s(s+1)(s+4)}$$

$$Y(s) = \frac{3}{s} - \frac{3-\beta-4\alpha}{s+1} + \frac{3-4\alpha-4\beta}{s+4} \quad \Rightarrow \quad y(t) = \left(\frac{3}{4} + \frac{-3+\beta+4\alpha}{3} e^{-t} + \frac{3-4\alpha-4\beta}{12} e^{-4t} \right) 1(t)$$

$u(t)$
↓
○

EXAMPLE 3.15 Forced Equation Solution with Zero Initial Conditions

$$\ddot{y}(t) + 5\dot{y}(t) + 4y(t) = u(t), \quad y(0) = 0, \quad \dot{y}(0) = 0, \quad u(t) = 2e^{-2t} 1(t)$$

$$s^2 Y(s) + 5sY(s) + 4Y(s) = \frac{2}{s+2} \quad Y(s) = \frac{2}{(s+2)(s+1)(s+4)}$$

$$Y(s) = -\frac{1}{s+2} + \frac{\frac{2}{3}}{s+1} + \frac{\frac{1}{3}}{s+4} \quad \Rightarrow \quad y(t) = \left(-1e^{-2t} + \frac{2}{3}e^{-t} + \frac{1}{3}e^{-4t} \right) 1(t)$$

$u(t)$
↓
○

Lecture 3- Review Laplace

53

53

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Homework

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$$G(s) = \frac{5s+3}{(s+1)(s+2)(s+3)} = \frac{5s+3}{s^3+6s^2+11s+6}$$

$$G(s) = \frac{K_{-1}}{s+1} + \frac{K_{-2}}{s+2} + \frac{K_{-3}}{s+3}$$

$$K_{-1} = [(s+1)G(s)] \Big|_{s=-1} = \frac{5(-1)+3}{(2-1)(3-1)} = -1$$

$$K_{-2} = [(s+2)G(s)] \Big|_{s=-2} = \frac{5(-2)+3}{(1-2)(3-2)} = 7$$

$$K_{-3} = [(s+3)G(s)] \Big|_{s=-3} = \frac{5(-3)+3}{(1-3)(2-3)} = -6$$

$$G(s) = \frac{-1}{s+1} + \frac{7}{s+2} - \frac{6}{s+3}$$

$$g(t) =$$

Lecture 3- Review Laplace

54

54

$$G(s) = \frac{1}{s(s+1)^3(s+2)} = \frac{1}{s^5 + 5s^4 + 9s^3 + 7s^2 + 2s}$$

Homework

$$G(s) = \frac{K_0}{s} + \frac{K_{-2}}{s+2} + \frac{A_1}{s+1} + \frac{A_2}{(s+1)^2} + \frac{A_3}{(s+1)^3}$$

$$K_0 = [sG(s)] \Big|_{s=0} = \frac{1}{2}$$

$$K_{-2} = [(s+2)G(s)] \Big|_{s=-2} = \frac{1}{2}$$

$$G(s) = \frac{1}{2s} + \frac{1}{2(s+2)} - \frac{1}{s+1} - \frac{1}{(s+1)^3}$$

 $g(t) =$

$$A_3 = [(s+1)^3 G(s)] \Big|_{s=-1} = -1$$

$$A_2 = \frac{d}{ds} [(s+1)^3 G(s)] \Big|_{s=-1} = \frac{d}{ds} \left[\frac{1}{s(s+2)} \right] \Big|_{s=-1} = 0$$

$$A_1 = \frac{1}{2} \frac{d^2}{ds^2} [(s+1)^3 G(s)] \Big|_{s=-1} = \frac{1}{2} \frac{d^2}{ds^2} \left[\frac{1}{s(s+2)} \right] \Big|_{s=-1} = -1$$

EXAMPLE 3.5

Frequency Response

~~延後~~

3.1.9 Linear System Analysis Using MATLAB

```

numb=[0 0 100]; % form numerator
denb=[1 10.1 101]; % form denominator
sysb=tf(numb,denb); % define system by its numerator and denominator
t=0:0.01:5; % form time vector
y=step(sysb,t) % compute step response;
plot(t,y) % plot step response

```

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3 Dynamic Response

3.1 Review of Laplace Transforms

3.2.1 The Block Diagram

3.3 Effect of Pole Locations

3.4 Time-Domain Specifications

3.5 Effects of Zeros and Additional Poles

3.6.3 Routh's Stability Criterion

1. To study **block diagrams**, their components, and their underlying mathematics.
2. To obtain **transfer function** of systems through block diagram manipulation and reduction.

Lecture 3 57

57

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Block-Diagram Elements of Comparators

(a) Subtraction

(b) Addition

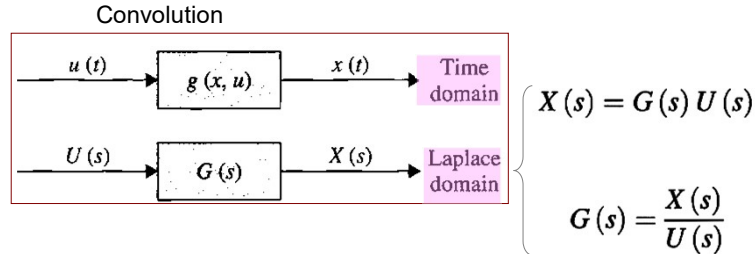
(c) Addition and subtraction

Block diagram elements of control systems.

Lecture 3 58

58

Time & Laplace Domain Block Diagrams

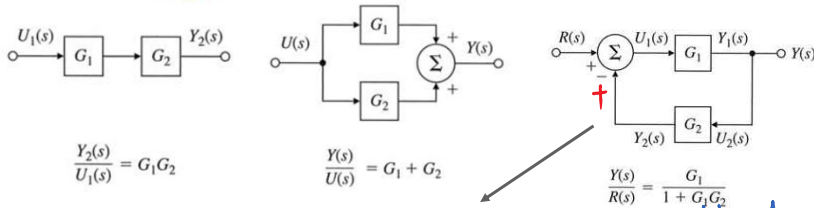


Convolution

$$x(t) = \int_0^{\infty} u(\tau)g(t-\tau)d\tau$$

3.2 System Modeling Diagrams

3.2.1 The Block Diagram



G_1 : gain 增益

negative feedback

Feedback Control 反馈控制

$$\begin{cases} U_1(s) = R(s) - Y_2(s), \\ Y_2(s) = G_2(s)G_1(s)U_1(s) \\ Y_1(s) = G_1(s)U_1(s), \end{cases}$$

$$U_1 = R - G_2 G_1 U_1$$

$$R = (1 + G_2 G_1) U_1$$

$$U_1 = \frac{1}{1 + G_2 G_1} R \Rightarrow Y_1(s) = \frac{G_1(s)}{1 + G_1(s)G_2(s)} R(s)$$

The gain of a single-loop negative feedback system is given by the forward gain divided by the sum of 1 plus the loop gain.

Negative feedback: $Y_1(s) = \frac{G_1(s)}{1 + G_1(s)G_2(s)} R(s)$

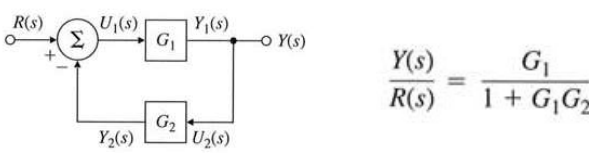
Positive feedback: $Y_1(s) = \frac{G_1(s)}{1 - G_1(s)G_2(s)} R(s)$

unity feedback system $G_2=1$

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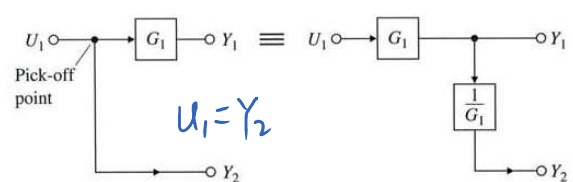
Rules for Block Diagram Manipulation

Rule 1:



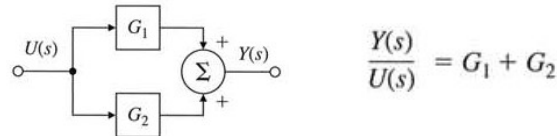
$$\frac{Y(s)}{R(s)} = \frac{G_1}{1 + G_1 G_2}$$

Rule 2:



$U_1 = Y_2$

Rule 3:



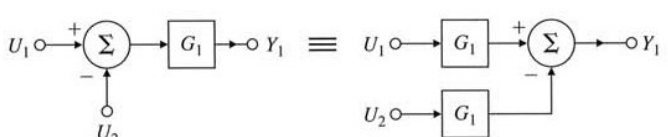
$$\frac{Y(s)}{U(s)} = G_1 + G_2$$

Lecture 3 61

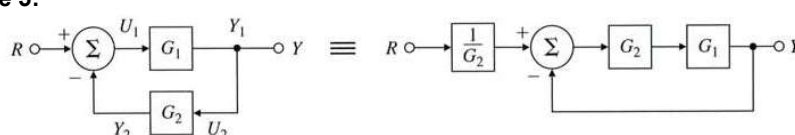
61

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Rule 4:



Rule 5:



$$\frac{Y}{R} = \frac{G_1}{1 + G_1 G_2}$$

$$\frac{Y}{R \cdot \frac{1}{G_2}} = \frac{G_1 G_2}{1 + G_1 G_2}$$

Lecture 3 62

62

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3.2.1 The Block Diagram

Terms

$r(t)$: reference input or command signal
 $y(t)$: sensed/actual output
 $e(t) = r(t) - ybar(t)$
 $u(t)$: control input
 Plant P
 $ybar$: estimate of y
Purpose: find compensator/controller
 { - error is "small"
 - stability is "good"
 If $r(t)$ is a constant \rightarrow "regulator" problem
 If $r(t)$ = function of time \rightarrow "servo" or "tracking" problem

伺服 追踪

Lecture 3 63

63

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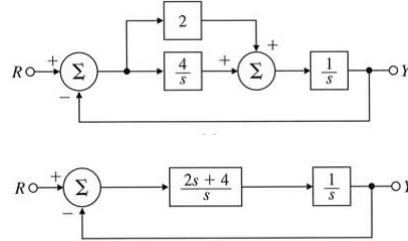
3.2.1 The Block Diagram

(1) Assume $d(t)=0$ $\frac{Y}{R} = \frac{G_c G_p}{1 + G_c \cdot G_p \cdot H}$
 (2) Assume $r(t)=0$ $\frac{Y}{D} = \frac{G_p}{1 + G_c \cdot G_p \cdot H}$

Lecture 3 64

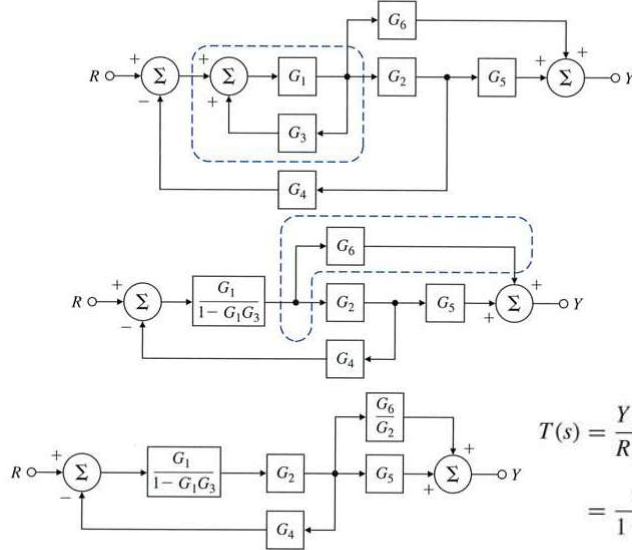
64

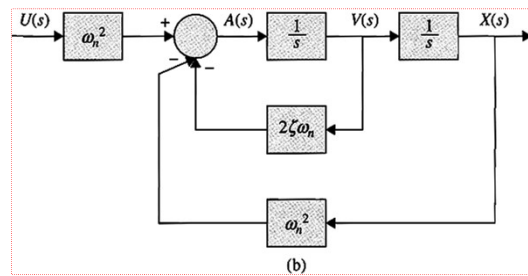
EXAMPLE 3.20 Transfer Function from a Simple Block Diagram



$$T(s) = \frac{Y(s)}{R(s)} = \frac{\frac{2s+4}{s^2}}{1 + \frac{2s+4}{s^2}} = \frac{2s+4}{s^2+2s+4}$$

EXAMPLE 3.21 Transfer Function from the Block Diagram





$$\frac{X(s)}{U(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

Mason's Rule for Complicated Block Diagram ✗
See Appendix W3.2.3 at www.fpe7e.com

3.2.2 Block Diagram Reduction Using MATLAB ✗